

## Lecture 16

### More z-Transform (Lathi 5.2,5.4-5.5)

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### Convolution property of z-transform

- If  $h[n]$  is the impulse response of a discrete-time LTI system, then

$$\text{then } H[z] = \sum_{n=-\infty}^{\infty} h[n]z^{-n} \quad h[n] \iff H[z]$$

- If  $x_1[n] \iff X_1[z]$  and  $x_2[n] \iff X_2[z]$ ,

$$\text{Then } x_1[n] * x_2[n] \iff X_1[z]X_2[z]$$

- That is: convolution in the time-domain is the same as multiplication in the z-domain.
- Therefore, we can derive the input-output relationship for any LTI systems in z-domain:

$$y[n] = x[n] * h[n]$$

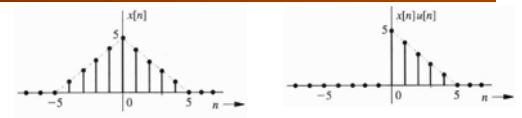
$$Y[z] = X[z]H[z]$$

L5.2 p511

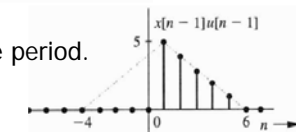
### Shift Property of z-Transform

- If  $x[n]u[n] \iff X[z]$

$$\text{then } x[n-1]u[n-1] \iff \frac{1}{z}X[z]$$

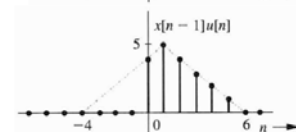


which is delay causal signal by 1 sample period.



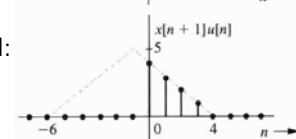
- If we delay  $x[n]$  first:

$$x[n-1]u[n] \iff \frac{1}{z}X[z] + x[-1]$$



- If we ADVANCE  $x[n]$  by 1 sample period:

$$x[n+1]u[n] \iff zX[z] - zx[0]$$



L5.2 p508

### More Properties of z-Transform

- For all these cases, we assume:  $x[n]u[n] \iff X[z]$

- Scaling Property:

$$\gamma^n x[n]u[n] \iff X\left[\frac{z}{\gamma}\right]$$

- Multiply by n property:

$$nx[n]u[n] \iff -z \frac{d}{dz} X[z]$$

- Time reversal property:

$$x[-n] \iff X[1/z]$$

- Initial value property:  $X[z] = \sum_{n=0}^{\infty} x[n]z^{-n} = x[0] + \frac{x[1]}{z} + \frac{x[2]}{z^2} + \frac{x[3]}{z^3} + \dots$

$$x[0] = \lim_{z \rightarrow \infty} X[z]$$

L5.2 p512

## Summary of z-transform properties (1)

Operation	$x[n]$	$X[z]$
Addition	$x_1[n] + x_2[n]$	$X_1[z] + X_2[z]$
Scalar multiplication	$ax[n]$	$aX[z]$
Right-shifting	$x[n - m]u[n - m]$	$\frac{1}{z^m} X[z]$
	$x[n - m]u[n]$	$\frac{1}{z^m} X[z] + \frac{1}{z^m} \sum_{n=1}^m x[-n]z^n$
	$x[n - 1]u[n]$	$\frac{1}{z} X[z] + x[-1]$
	$x[n - 2]u[n]$	$\frac{1}{z^2} X[z] + \frac{1}{z} x[-1] + x[-2]$
	$x[n - 3]u[n]$	$\frac{1}{z^3} X[z] + \frac{1}{z^2} x[-1] + \frac{1}{z} x[-2] + x[-3]$

L5.2 p514

## Summary of z-transform properties (2)

Operation	$x[n]$	$X[z]$
Left-shifting	$x[n + m]u[n]$	$z^m X[z] - z^m \sum_{n=0}^{m-1} x[n]z^{-n}$
	$x[n + 1]u[n]$	$zX[z] - zx[0]$
	$x[n + 2]u[n]$	$z^2 X[z] - z^2 x[0] - zx[1]$
Multiplication by $\gamma^n$	$x[n + 3]u[n]$	$z^3 X[z] - z^3 x[0] - z^2 x[1] - zx[2]$
	$\gamma^n x[n]u[n]$	$X\left[\frac{z}{\gamma}\right]$
	Multiplication by $n$	$n x[n]u[n]$
$x_1[n] * x_2[n]$		$X_1[z] X_2[z]$
Time convolution	$x[-n]$	$X[1/z]$
Time reversal	$x[0]$	$\lim_{z \rightarrow \infty} X[z]$
Initial value	$\lim_{N \rightarrow \infty} x[N]$	$\lim_{z \rightarrow 1} (z - 1) X[z]$ poles of $(z - 1) X[z]$ inside the unit circle

L5.2 p514

## Discrete LTI System and Difference Equation

- Consider a discrete time system where the input-output relation is described by:  $y[n] - 5y[n-1] + 6y[n-2] = x[n] + 3x[n-1] + 5x[n-2]$
- This is known as a difference equation, where current output is dependent on current input  $x[n]$ , and two previous inputs and outputs  $x[n-1]$ ,  $x[n-2]$ ,  $y[n-1]$  and  $y[n-2]$ .
- Take z-transform on both sides and assume zero-state condition:

$$Y(z) - 5z^{-1}Y(z) + 6z^{-2}Y(z) = X(z) + 3z^{-1}X(z) + 5z^{-2}X(z)$$

$$\Rightarrow (1 - 5z^{-1} + 6z^{-2})Y(z) = (1 + 3z^{-1} + 5z^{-2})X(z)$$

$$\Rightarrow \frac{Y(z)}{X(z)} = H(z) = \frac{1 + 3z^{-1} + 5z^{-2}}{1 - 5z^{-1} + 6z^{-2}}$$

- The transfer function of a general Nth order causal discrete LTI system is:

$$H[z] = \frac{b_0 + b_1 z^{-1} + \dots + b_{N-1} z^{-N+1} + b_N z^{-N}}{1 + a_1 z^{-1} + \dots + a_{N-1} z^{-N+1} + a_N z^{-N}}$$

L5.4 p525

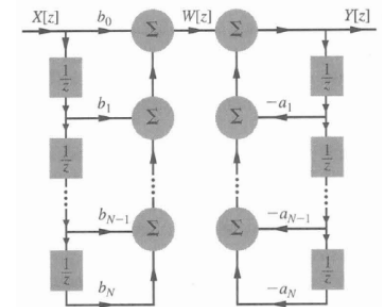
## Realization of LTI System – Direct Form I

- The general transfer function  $H(z)$  can be realised using Direct Form I as follows:

$$Y[z] = H[z]X[z] = \frac{b_0 + b_1 z^{-1} + \dots + b_{N-1} z^{-N+1} + b_N z^{-N}}{1 + a_1 z^{-1} + \dots + a_{N-1} z^{-N+1} + a_N z^{-N}} X[z]$$

$$= \left( \frac{1}{1 + a_1 z^{-1} + \dots + a_{N-1} z^{-N+1} + a_N z^{-N}} \right) (b_0 + b_1 z^{-1} + \dots + b_{N-1} z^{-N+1} + b_N z^{-N}) X[z]$$

$$= \left( \frac{1}{1 + a_1 z^{-1} + \dots + a_{N-1} z^{-N+1} + a_N z^{-N}} \right) W(z)$$



L5.4 p525

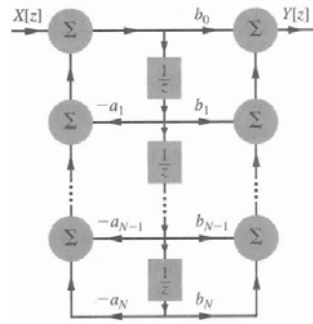
## Realization of LTI System – Direct Form II

- Or use Canonical Director Form II:

$$Y[z] = H[z]X[z] = \frac{b_0 + b_1z^{-1} + \dots + b_{N-1}z^{-N+1} + b_Nz^{-N}}{1 + a_1z^{-1} + \dots + a_{N-1}z^{-N+1} + a_Nz^{-N}} X[z]$$

$$= (b_0 + b_1z^{-1} + \dots + b_{N-1}z^{-N+1} + b_Nz^{-N}) \left( \frac{1}{1 + a_1z^{-1} + \dots + a_{N-1}z^{-N+1} + a_Nz^{-N}} \right) X[z]$$

$$= (b_0 + b_1z^{-1} + \dots + b_{N-1}z^{-N+1} + b_Nz^{-N}) W(z)$$



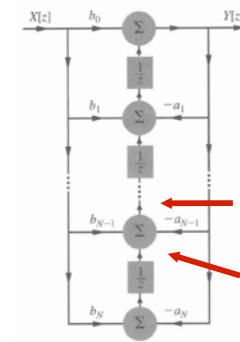
L5.4 p525

## Realization of LTI System – Transposed Direct Form II

- Or the transposed version:  $Y[z] = H[z]X[z] = \frac{b_0 + b_1z^{-1} + \dots + b_{N-1}z^{-N+1} + b_Nz^{-N}}{1 + a_1z^{-1} + \dots + a_{N-1}z^{-N+1} + a_Nz^{-N}} X[z]$

$$Y[z] + (a_1z^{-1} + \dots + a_{N-1}z^{-N+1} + a_Nz^{-N})Y[z] = (b_0 + b_1z^{-1} + \dots + b_{N-1}z^{-N+1} + b_Nz^{-N})X[z]$$

$$Y[z] = (b_0 + b_1z^{-1} + \dots + b_{N-1}z^{-N+1} + b_Nz^{-N})X[z] - (a_1z^{-1} + \dots + a_{N-1}z^{-N+1} + a_Nz^{-N})Y[z]$$



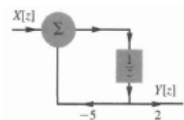
$$(b_N X[z] - a_N Y[z])z^{-1} + (b_{N-1} X[z] - a_{N-1} Y[z])$$

$$(b_N X[z] - a_N Y[z])z^{-1}$$

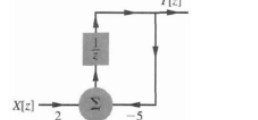
L5.4 p525

## Examples

Direct Form II

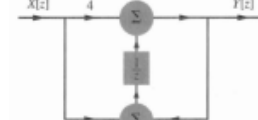
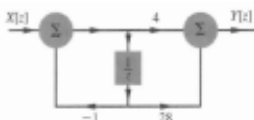


Transposed

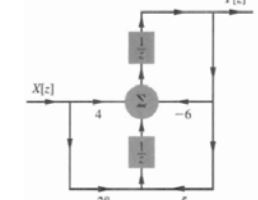
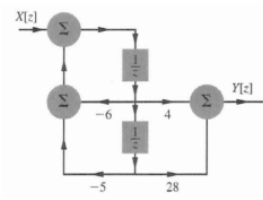


$$H[z] = \frac{2}{z+5} = \frac{2z^{-1}}{1+5z^{-1}}$$

$$H[z] = \frac{4z+28}{z+1} = \frac{4+28z^{-1}}{1+z^{-1}}$$



$$H[z] = \frac{4z+28}{z^2+6z+5} = \frac{4+28z^{-1}}{1+6z^{-1}+5z^{-2}}$$

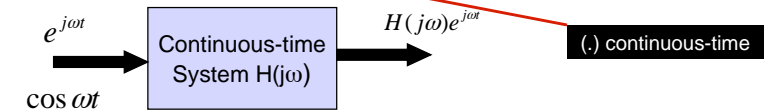


L5.4 p527

## Frequency Response of Discrete-time Systems

- Remember that for a continuous-time system, the system response to an input  $e^{j\omega t}$  is  $H(j\omega) e^{j\omega t}$ .

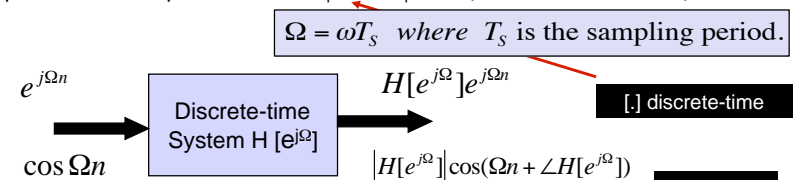
- The response to an input  $\cos \omega t$  is  $|H(j\omega)| \cos(\omega t + \angle H(j\omega))$ .



- Now, consider a discrete-time system with z-domain transfer function  $H[z]$ .

- Let  $z = e^{j\Omega}$ , the system response to an input  $e^{j\Omega n}$  is  $H[e^{j\Omega}] e^{j\Omega n}$ .

- The response to an input  $\cos \Omega n$  is  $|H[e^{j\Omega}]| \cos(\cos \Omega n + \angle H[e^{j\Omega}])$ .



$$\Omega = \omega T_s \text{ where } T_s \text{ is the sampling period.}$$

L5.5 p531

## Frequency Response Example (1)

- For a system specified by the following difference equation, find the frequency response of the system.

$$y[n+1] - 0.8y[n] = x[n+1]$$

- Take z-transform on both sides to find the transfer function:

$$zY[z] - 0.8Y[z] = zX[z] \Rightarrow H[z] = \frac{Y[z]}{X[z]} = \frac{z}{z - 0.8} = \frac{1}{1 - 0.8z^{-1}}$$

- Therefore the frequency response is:

$$H[e^{j\Omega}] = \frac{1}{1 - 0.8e^{-j\Omega}} = \frac{1}{(1 - 0.8 \cos \Omega) + j0.8 \sin \Omega}$$

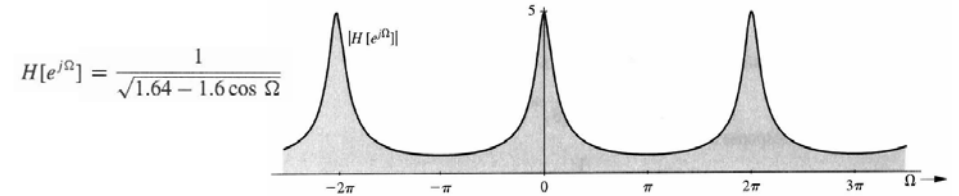
$$|H[e^{j\Omega}]| = \frac{1}{\sqrt{(1 - 0.8 \cos \Omega)^2 + (0.8 \sin \Omega)^2}} = \frac{1}{\sqrt{1.64 - 1.6 \cos \Omega}}$$

$$\angle H[e^{j\Omega}] = -\tan^{-1} \left[ \frac{0.8 \sin \Omega}{1 - 0.8 \cos \Omega} \right]$$

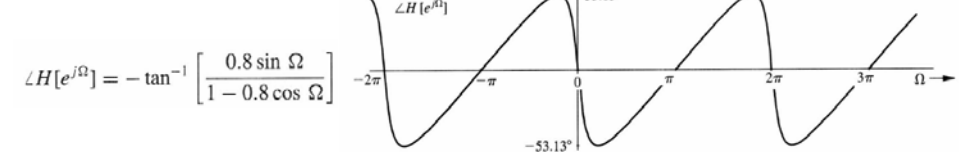
L5.5 p533

## Frequency Response Example (2)

- Amplitude response



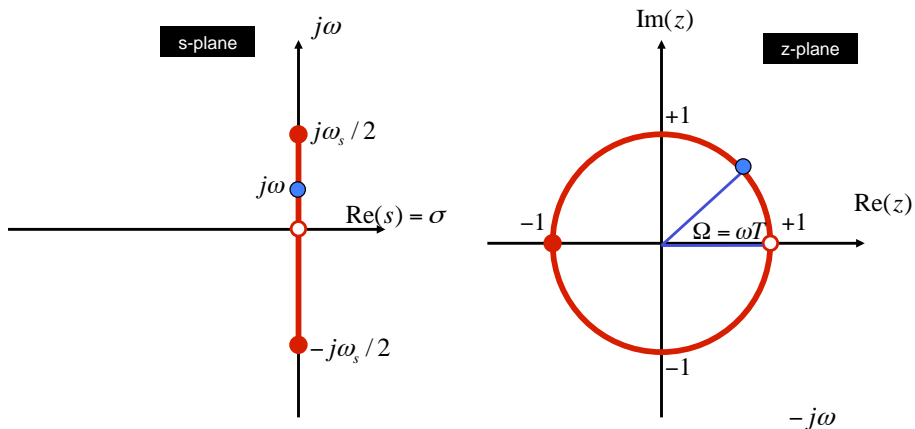
- Phase response



L5.5 p533

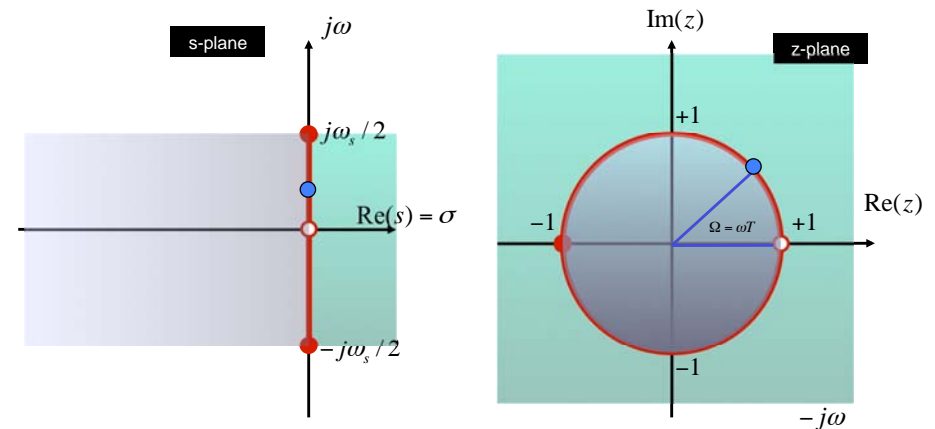
## Mapping from s-plane to z-plane

- Since  $z = e^{sT} = e^{(\sigma + j\omega)T} = e^{\sigma T} e^{j\omega T}$  where  $T = 2\pi/\omega_s$  we can map the s-plane to the z-plane as below:



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